

Power Electronics

Lecture 3

BOOST and BUCK-BOOST DC-DC converters

01

BOOST converter

02

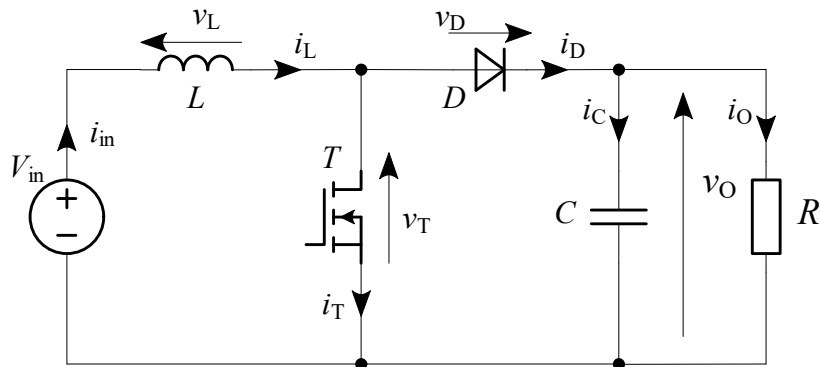
BUCK-BOOST converter

03

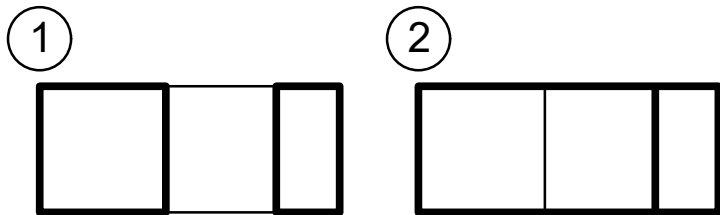
SEPIC and other BUCK-BOOST converters

Boost converter

The boost converter (a step-up converter) is composed of a transistor, a diode, a single inductor and capacitor.

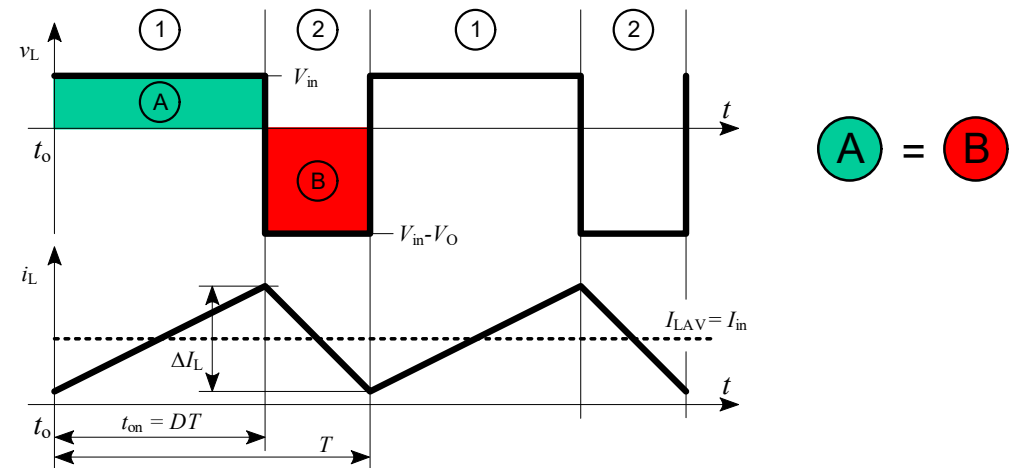


Transistor T is switched between the on- and off-states. For CCM there are two equivalent circuits



Inductor voltage is rectangular and equal to V_{in} when transistor is on, and $V_{in} - V_O$ when transistor is in off-state.

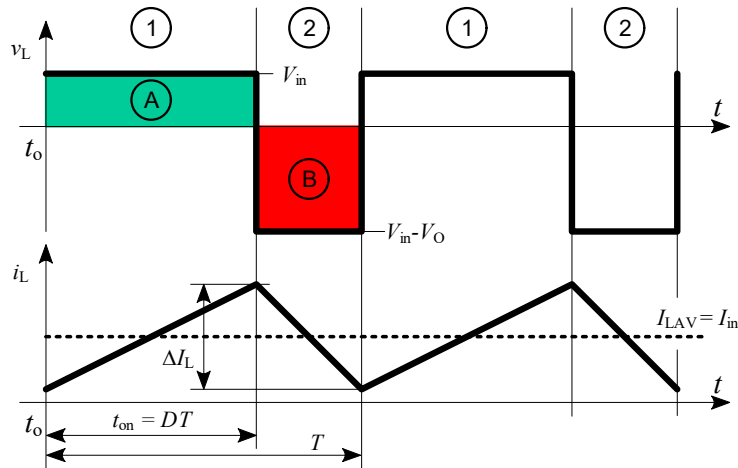
In the steady state operation:



$$v_L(t) = L \frac{di(t)}{dt} \quad \Delta I_L = \frac{1}{L} \int_{t_o}^{t_o+DT} V_{in} dt = \frac{V_{in}}{L} DT$$

The output voltage

Due to the fact that in a steady state the average value of the inductor voltage is zero it is possible to derive the output voltage as a function of D and input voltage V_{in} .



$$V_{in}DT + (V_{in} - V_O)(1 - D)T = 0$$

$$V_{in}D + V_{in} - V_O - V_{in}D + V_OD = 0$$

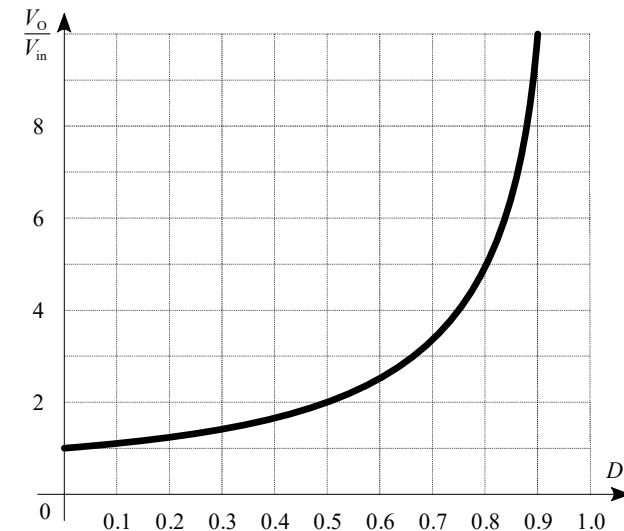
$$V_{in} - V_O(1 - D) = 0$$

$$V_O = \frac{V_{in}}{1 - D}$$

The characteristic $V_O = V_{in}/(1 - D)$ is valid only when the converter works in CCM

The output voltage control characteristic

$$\frac{V_O}{V_{in}} = \frac{1}{1 - D}$$



CCM operation

Inductor current ripple is given as:

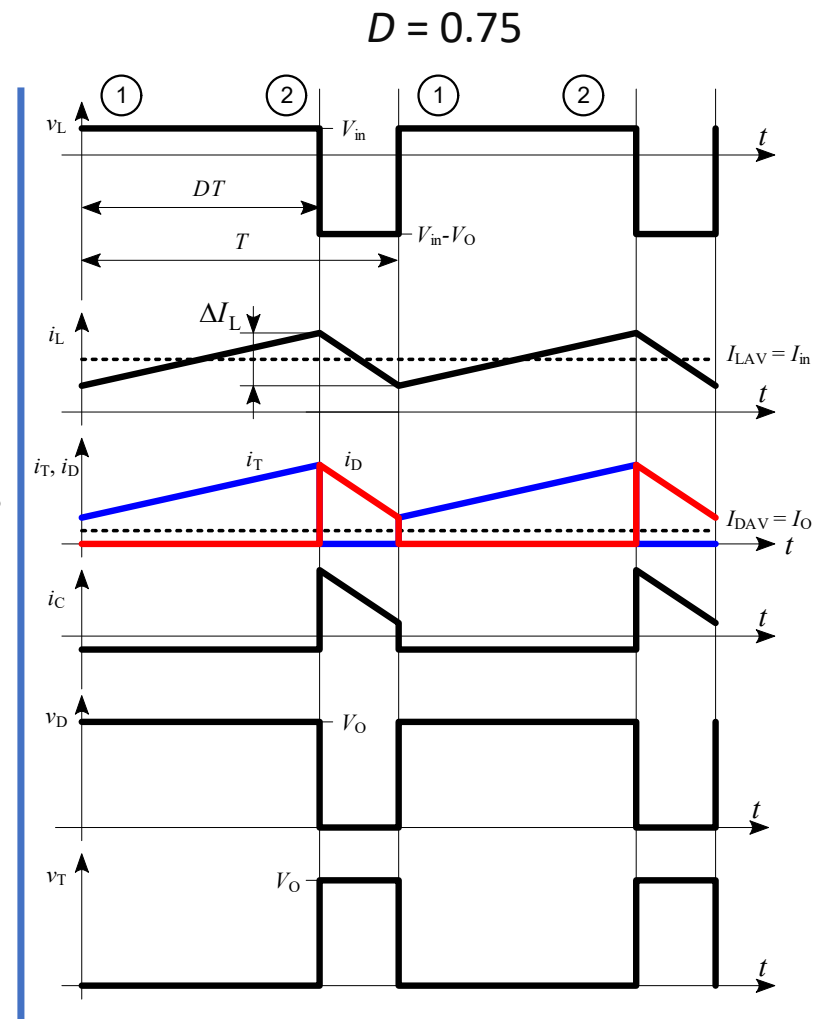
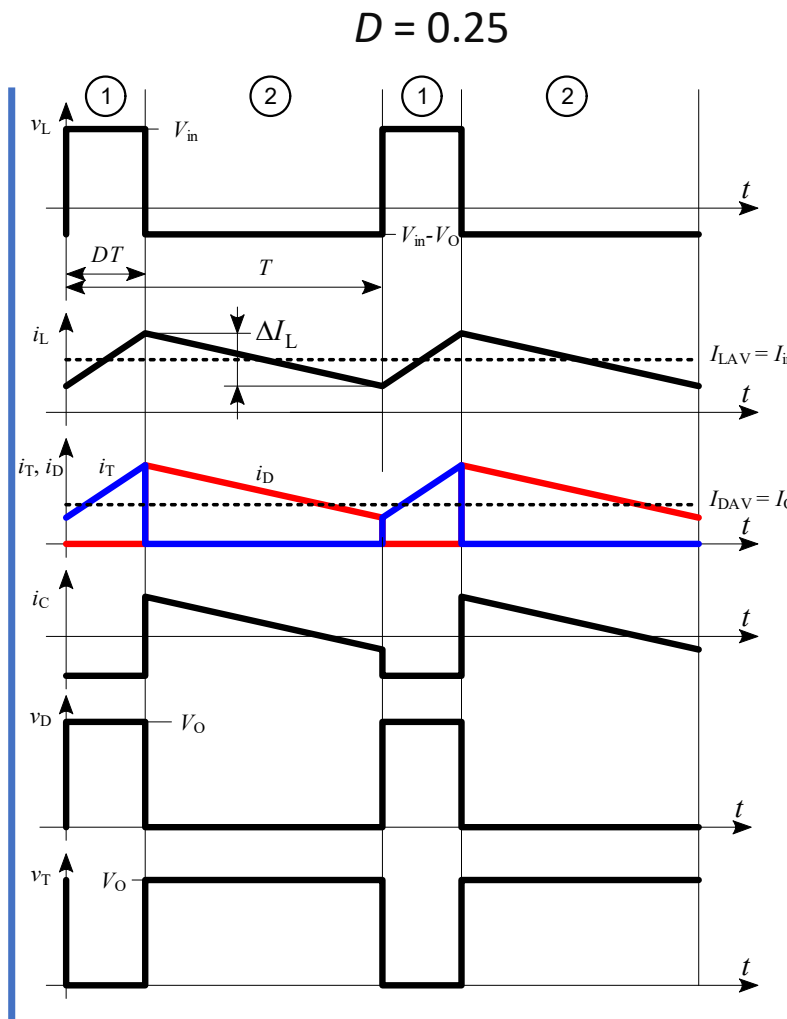
$$\Delta I_L = \frac{1}{L} \int_0^{DT} v_L(t) dt = \frac{V_{in}}{L} DT$$

Average value of the inductor current is equal to the average input current:

$$I_{LAV} = I_{in}$$

When it is assumed that converter is lossless

$$I_{LAV} = I_{in} \stackrel{P_{in}=P_O}{=} I_O \frac{V_O}{V_{in}} = I_O \frac{1}{1-D}$$

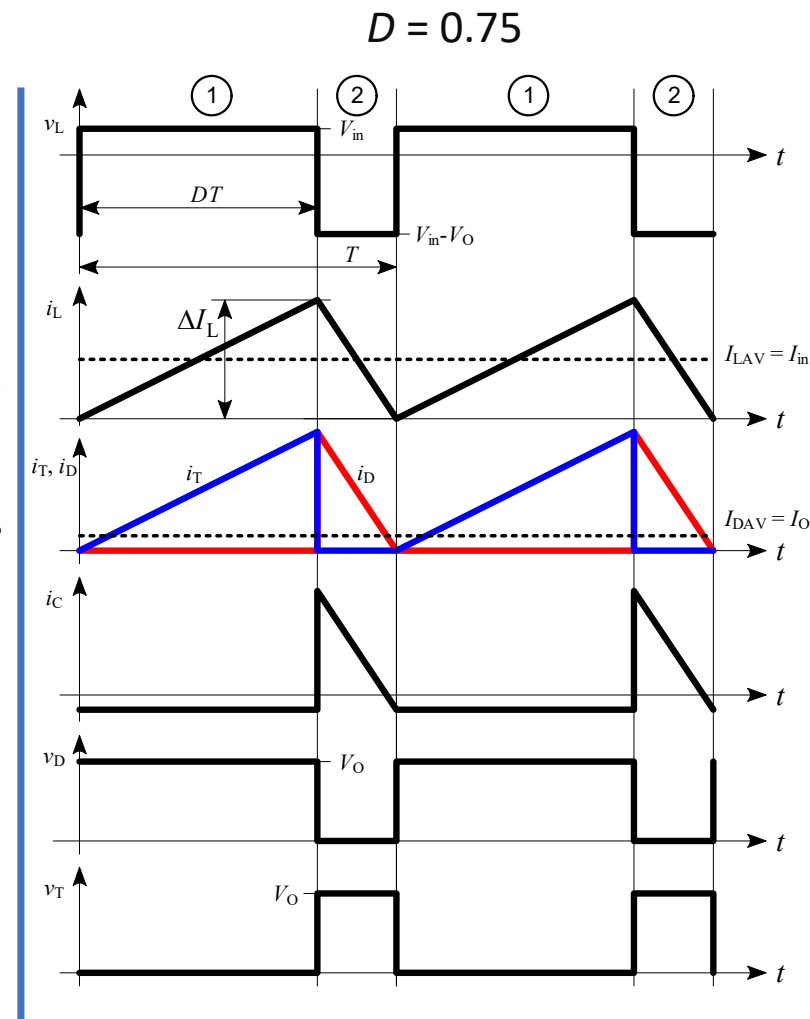
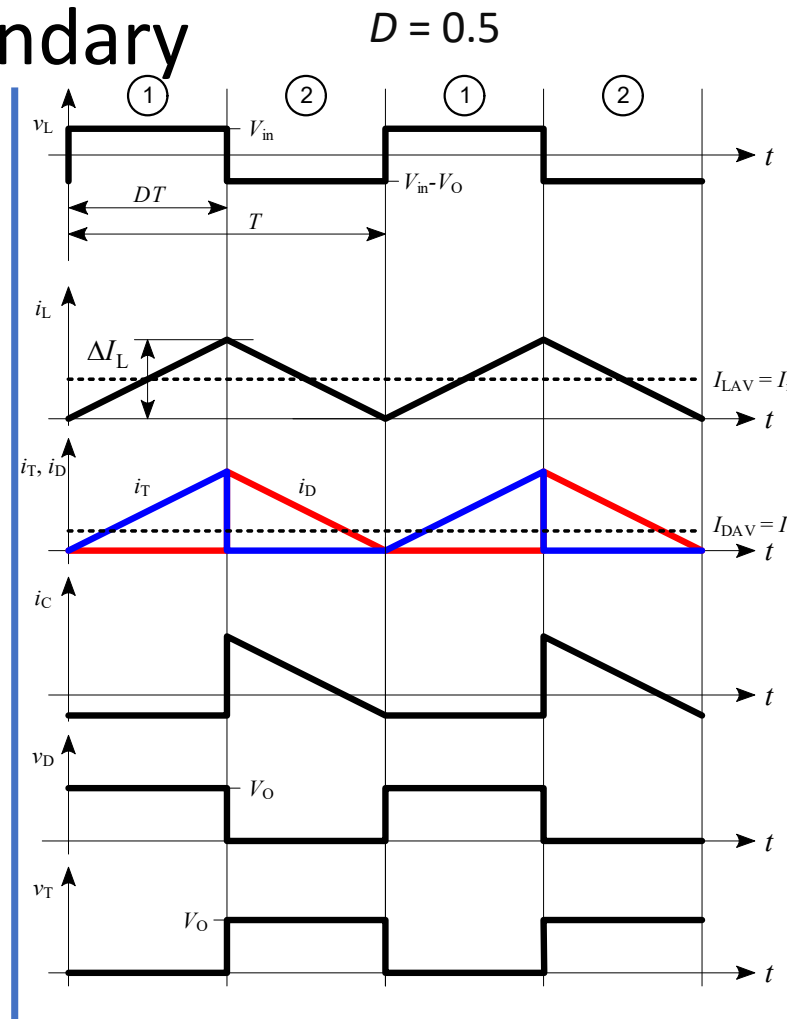
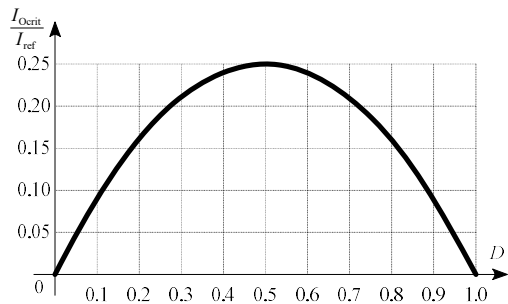


CCM/DCM boundary

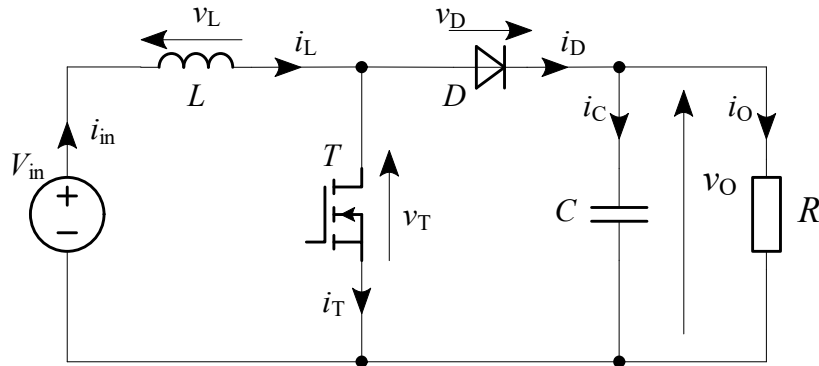
When the output current $I_O = I_{Ocrit} = \frac{1}{2}\Delta I_L(1-D)$ the converter operates at boundary between CCM and DCM

$$I_{Ocrit} = \frac{1}{2}\Delta I_L(1-D) = \frac{V_{in}T}{2L}D(1-D)$$

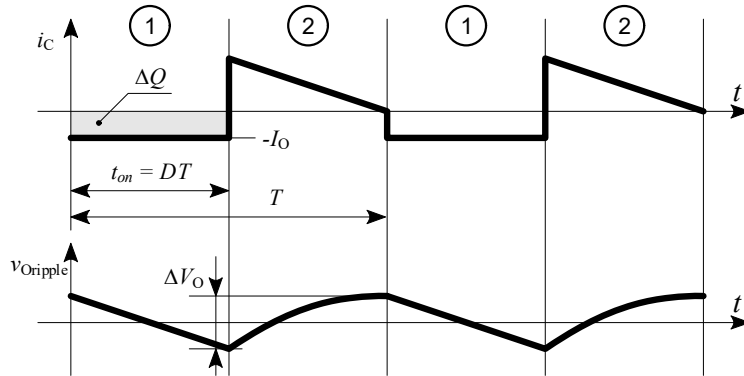
$$I_{Ocrit} = I_{ref}D(1-D)$$



Capacitor current and voltage



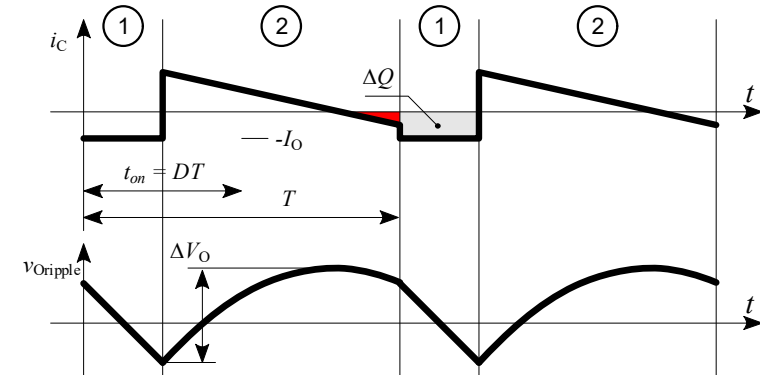
The capacitor current is an ac component of the diode current.



When $I_O D > \frac{1}{2} \Delta I_L (1-D)$ (in CCM) the charge ΔQ during the first equivalent circuit is:

$$\Delta Q = DT \cdot I_O \quad \Delta V_O = \frac{\Delta Q}{C} = \frac{DT I_O}{C}$$

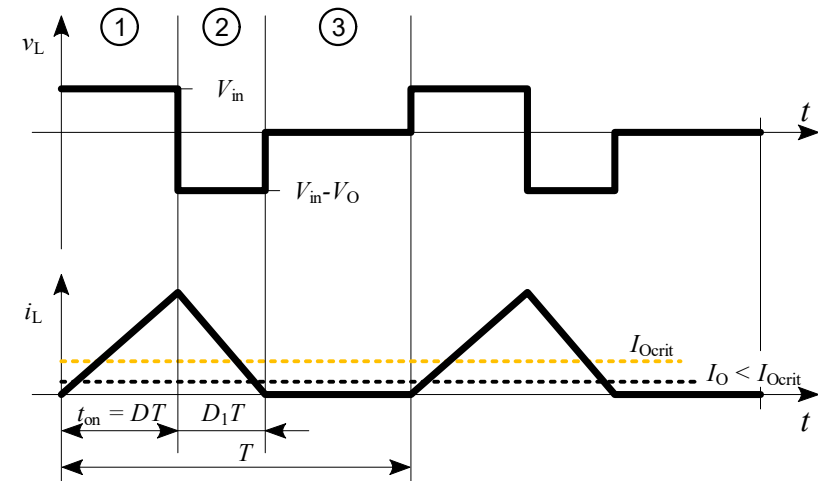
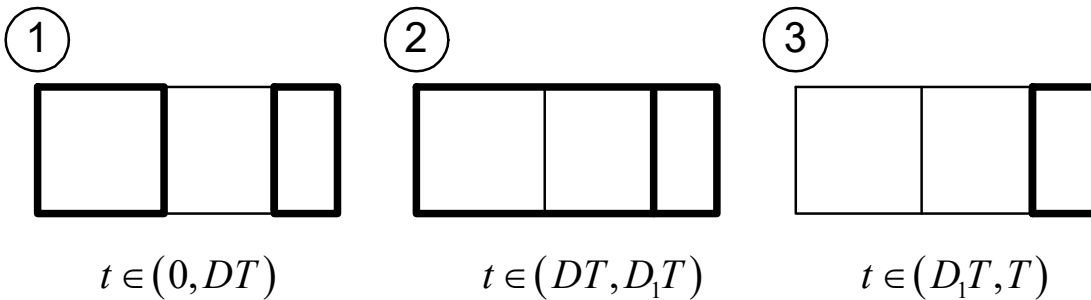
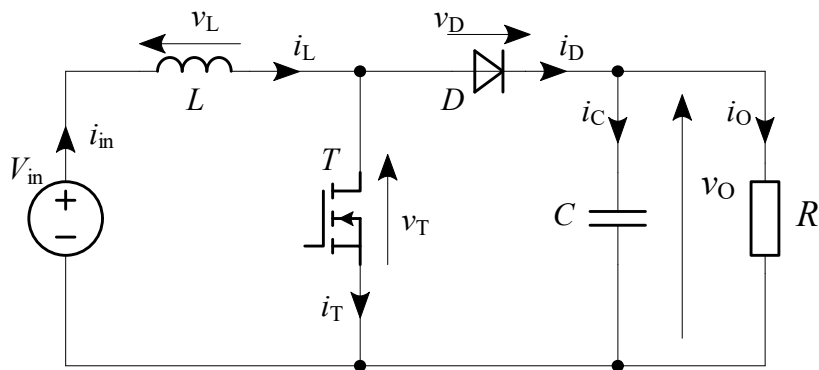
When $I_O D < \frac{1}{2} \Delta I_L (1-D)$ (in CCM) a small portion of the charge also occur at the second equivalent circuit and $\Delta Q > DT I_O$



Due to simplification of the analysis, it is assumed that $\Delta Q = DT I_O$

DCM operation

When the output current $I_O < I_{Ocrit}$ the converter operates in discontinuous conduction mode (DCM)



$$V_{LAV} = V_{in}D + (V_{in} - V_O)D_1 = 0$$

$$D_1 = \frac{V_{in}}{V_O - V_{in}} D$$

$$I_{in} = \frac{1}{2} \Delta I_L (D + D_1), \text{ where } \Delta I_L = \frac{V_{in}}{L} DT$$

$$I_O = \frac{V_{in}}{V_O} I_{in} = \frac{V_{in}}{V_O} \frac{1}{2} \frac{V_{in}}{L} DT (D + D_1) = \frac{V_{in}}{2L V_O} D^2 T \left(1 + \frac{V_{in}}{V_O - V_{in}} \right)$$

$$I_O = \frac{V_{in}}{2L V_O} TD^2 \left(\frac{V_O}{V_O - V_{in}} \right)^{I_{ref} = \frac{V_{in}T}{2L}} = I_{ref} \left(\frac{V_{in}}{V_O - V_{in}} \right) D^2$$

Output characteristics

In DCM

$$I_O = I_{\text{ref}} \left(\frac{V_{\text{in}}}{V_O - V_{\text{in}}} \right) D^2$$

$$\frac{I_O}{I_{\text{ref}}} \frac{1}{D^2} = \left(\frac{V_{\text{in}}}{V_O - V_{\text{in}}} \right) \Rightarrow \frac{V_O}{V_{\text{in}}} - 1 = \frac{I_O}{I_{\text{ref}}} \frac{1}{D^2}$$

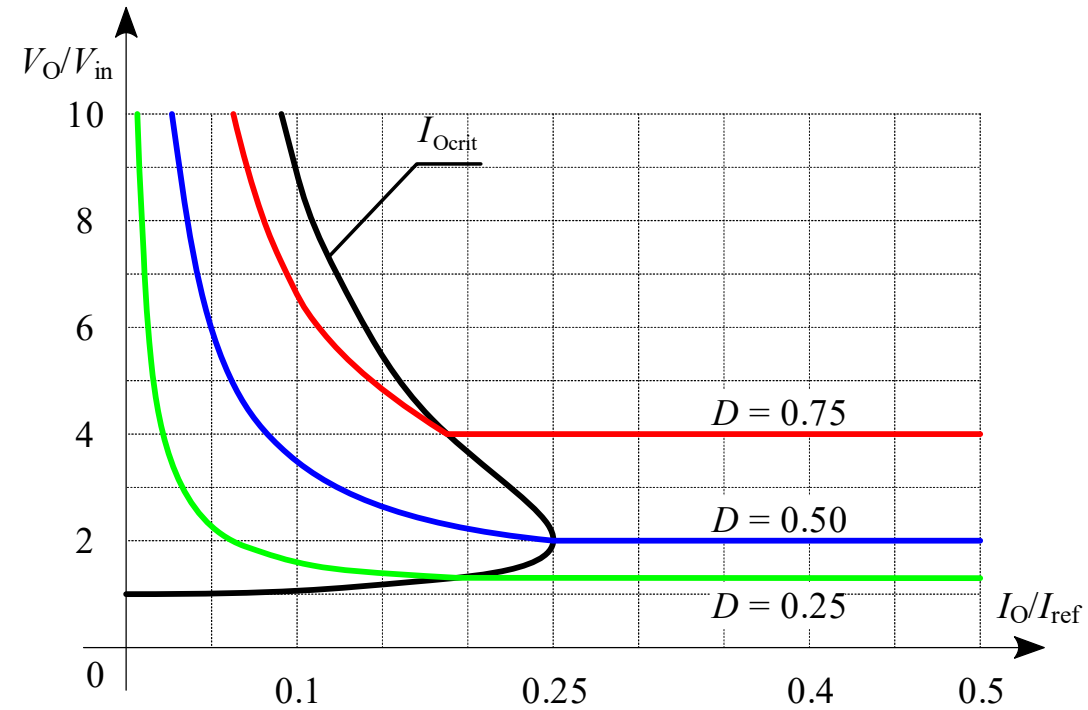
$$\frac{V_O}{V_{\text{in}}} = \frac{I_O}{I_{\text{ref}}} \frac{1}{D^2} + 1$$

In CCM

$$\frac{V_O}{V_{\text{in}}} = \frac{1}{1-D}$$

At CCM/DCM boundary

$$I_{\text{Ocrit}} = I_{\text{ref}} D(1-D) \quad D = \frac{V_O - V_{\text{in}}}{V_O} \quad I_{\text{Ocrit}} = I_{\text{ref}} \frac{V_O - V_{\text{in}}}{V_O} \left(1 - \frac{V_O - V_{\text{in}}}{V_O} \right) = I_{\text{ref}} \frac{\frac{V_O}{V_{\text{in}}} - 1}{\left(\frac{V_O}{V_{\text{in}}} \right)^2}$$



Boost with $V_O = \text{const}$

$$I_O = \frac{V_{in}}{2L} \frac{V_{in}}{V_O} T D^2 \left(\frac{V_O}{V_O - V_{in}} \right)^{I_{Oref} = \frac{V_O T}{2L}} = I_{Oref} \frac{V_{in}}{V_O} \left(\frac{V_{in}}{V_O - V_{in}} \right) D^2$$

In DCM

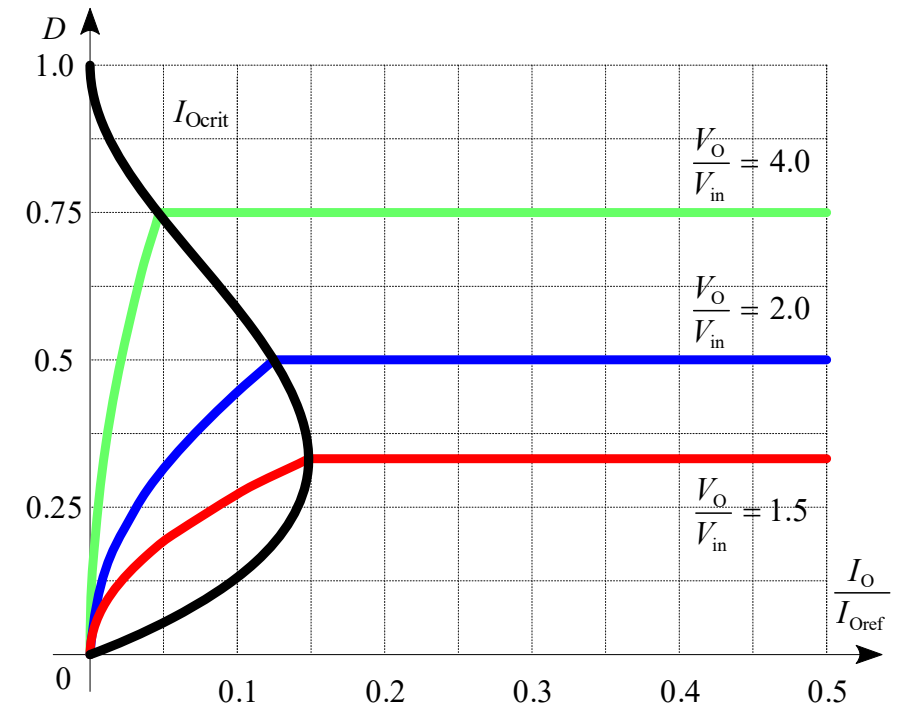
$$D^2 = \frac{\frac{I_O}{I_{Oref}}}{\frac{V_{in}}{V_O} \left(\frac{V_{in}}{V_O - V_{in}} \right)} \Rightarrow D = \sqrt{\frac{\frac{I_O}{I_{Oref}}}{\frac{V_{in}}{V_O} \left(\frac{V_{in}}{V_O - V_{in}} \right)}}$$

In CCM $D = 1 - \frac{V_{in}}{V_O}$

At CCM/DCM boundary

$$I_{Ocrit} = \frac{V_{in} T}{2L} \frac{V_O}{V_O} D(1-D)$$

$$I_{Ocrit} = \frac{I_{Oref} = \frac{V_O T}{2L}}{I_{Oref}} \frac{V_{in}}{V_O} D(1-D) = I_{Oref} (1-D) D(1-D) = I_{Oref} D(1-D)^2$$



Boost converter example #1

Boost converter example as a MPPT converter in solar systems.

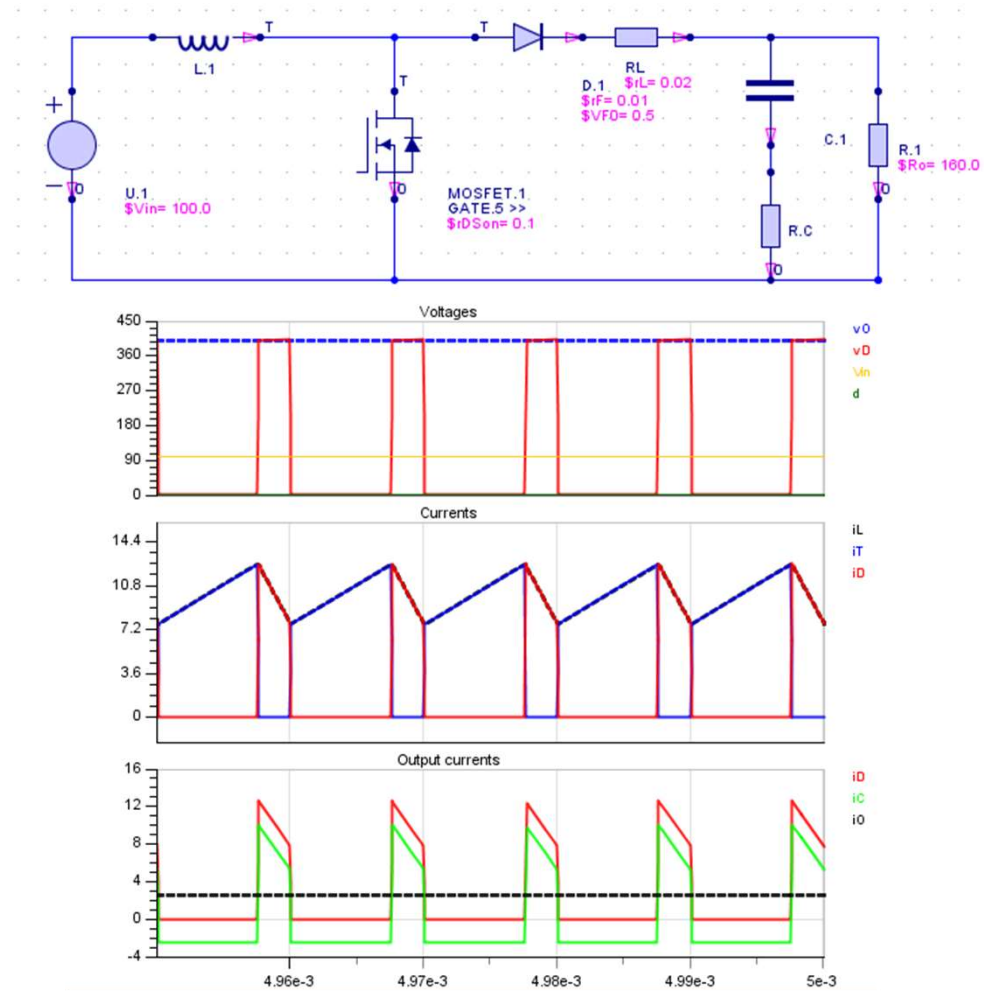
Input voltage: $V_{in} = 100 \text{ V}$,
 Output voltage: $V_{out} = 400 \text{ V}$,
 Switching frequency: $f = 100 \text{ kHz}$, ($T = 10 \mu\text{s}$)
 Output capacitance: $C = 10 \mu\text{F}$,
 Output power: $P_O = 1000 \text{ W}$,
 Inductor current pk-pk $\Delta I_L = 50\% I_{LAV}$.

$$I_O = \frac{P_O}{V_O} = \frac{1000 \text{ W}}{400 \text{ V}} = 2.5 \text{ A} \quad R = \frac{V_O}{I_O} = \frac{400 \text{ V}}{2.5 \text{ A}} = 160 \Omega$$

$$I_{in} = \frac{P_{in}}{V_{in}} = \frac{P_{in}=P_O}{100 \text{ V}} = 10 \text{ A} \quad D = 1 - \frac{V_{in}}{V_O} = 1 - \frac{100 \text{ V}}{400 \text{ V}} = 0.75$$

$$\Delta I_L = \frac{V_{in}}{L} DT \Rightarrow L = \frac{V_{in}}{\Delta I_L} DT = \frac{100 \text{ V}}{0.5 \cdot 10 \text{ A}} 0.75 \cdot 10 \mu\text{s} = 150 \mu\text{H}$$

GeckoCIRCUITS model of the boost converter



Boost converter example #2

Boost converter example in battery level voltage applications

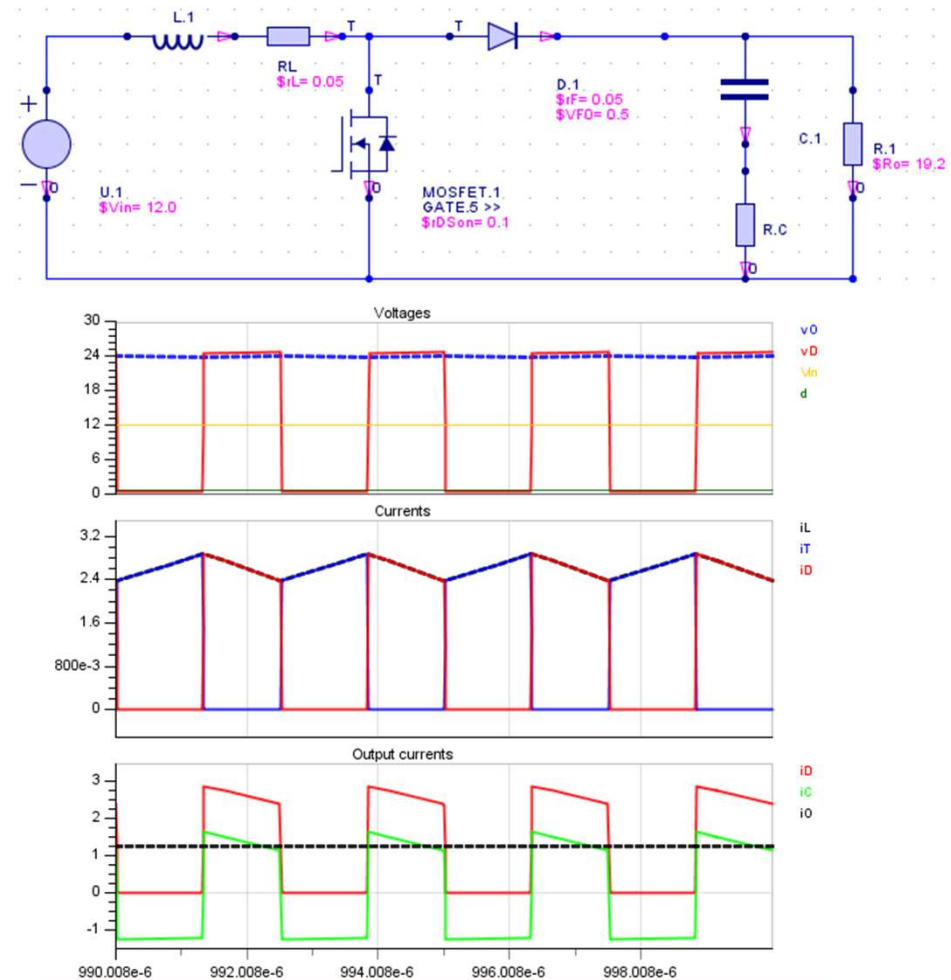
Input voltage: $V_{in} = 12 \text{ V}$,
 Output voltage: $V_{out} = 24 \text{ V}$,
 Switching frequency: $f = 400 \text{ kHz}$, ($T = 2.5 \text{ } \mu\text{s}$)
 Output capacitance: $C = 10 \text{ } \mu\text{F}$,
 Output power: $P_O = 30 \text{ W}$,
 Inductor current pk-pk $\Delta I_L = 20\% I_{LAV}$.

$$I_O = \frac{P_O}{V_O} = \frac{30 \text{ W}}{24 \text{ V}} = 1.25 \text{ A} \quad R = \frac{V_O}{I_O} = \frac{24 \text{ V}}{1.25 \text{ A}} = 19.2 \text{ } \Omega$$

$$I_{in} = \frac{P_{in}}{V_{in}} = \frac{P_{in}=P_O}{12 \text{ V}} = 2.5 \text{ A} \quad D = 1 - \frac{V_{in}}{V_O} = 1 - \frac{12 \text{ V}}{24 \text{ V}} = 0.5$$

$$\Delta I_L = \frac{V_{in}}{L} DT \Rightarrow L = \frac{V_{in}}{\Delta I_L} DT = \frac{12 \text{ V}}{0.2 \cdot 2.5 \text{ A}} 0.5 \cdot 2.5 \text{ } \mu\text{s} = 30 \text{ } \mu\text{H}$$

GeckoCIRCUITS model of the boost converter



Output voltage of other DC-DC converters

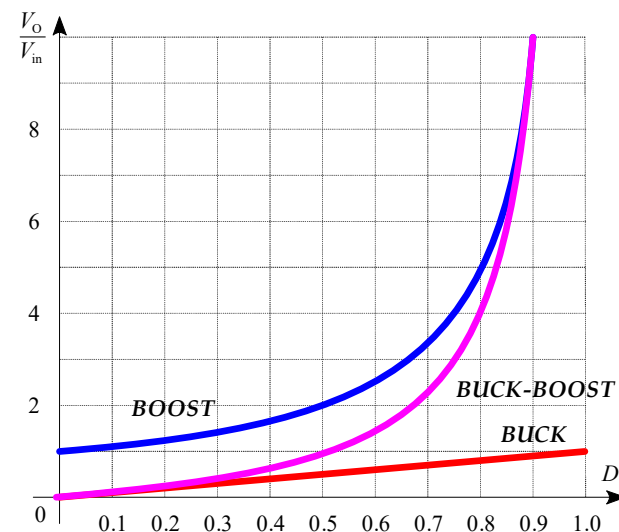
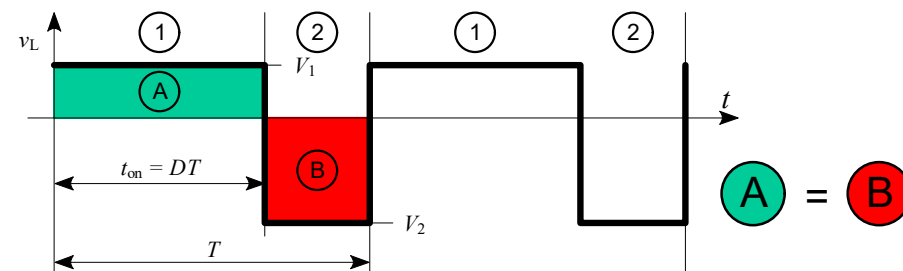
In basic DC-DC converter topologies voltage across the inductor can be equal to V_{in} , $-V_O$, $V_{in}-V_O$.

It is always true, that in the first equivalent circuit voltage is positive whereas this voltage is negative in the second equivalent circuit.

The possible combinations:

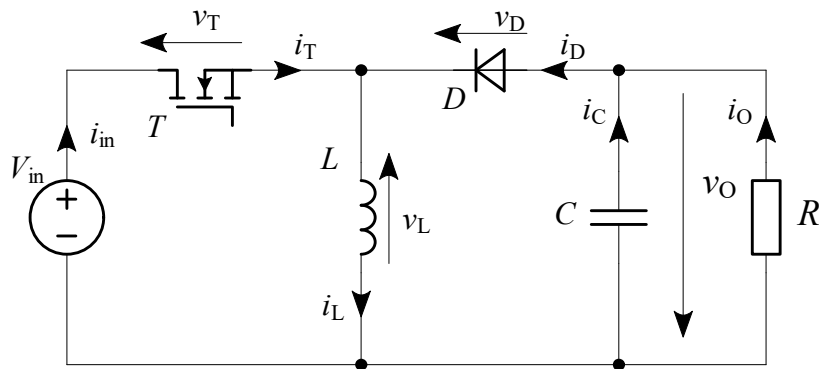
Topology	V_1	V_2	V_O/V_{in} (CCM)
Step down	$V_{in}-V_O$	$-V_O$	D
Step up	V_{in}	$V_{in}-V_O$	$\frac{1}{1-D}$
Step up-down	V_{in}	$-V_O$	$\frac{D}{1-D}$

Inductor voltage is rectangular

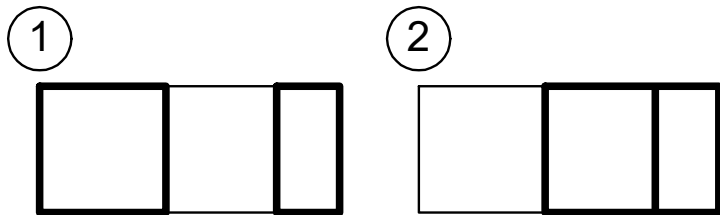


Buck-boost converter

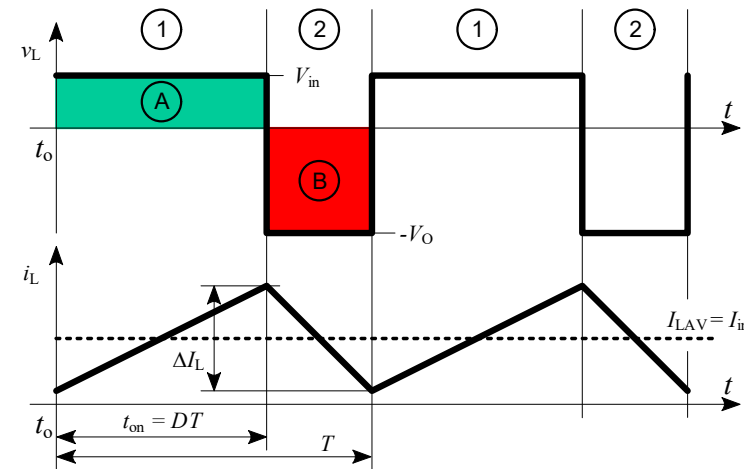
The buck-boost converter (a step up-down converter) has negative output voltage referred to negative terminal of the input source.



Transistor T is switched between the on- and off-states. For CCM there are two equivalent circuits



Average value of the inductor voltage in the steady state operation is equal to zero



$$\textcircled{A} = \textcircled{B}$$

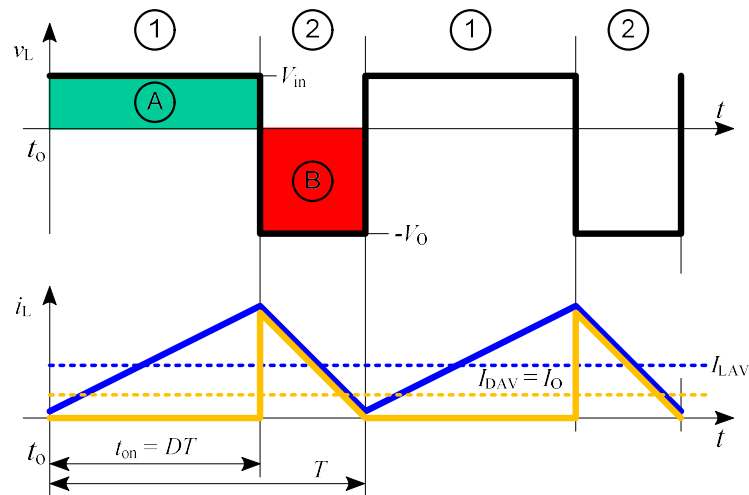
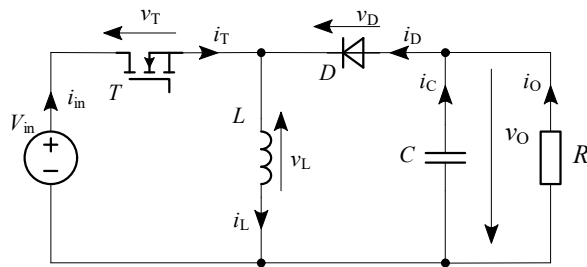
$$V_{in}DT + (-V_O)(1-D)T = 0$$

$$V_{in}D - V_O(1-D) = 0$$

$$V_O = V_{in} \frac{D}{1-D}$$

Buck-boost converter

Inductor current average value is not equal to the average value of the output current as in buck converter.



Critical inductor current I_{Lcrit} is at the boundary between CCM & DCM

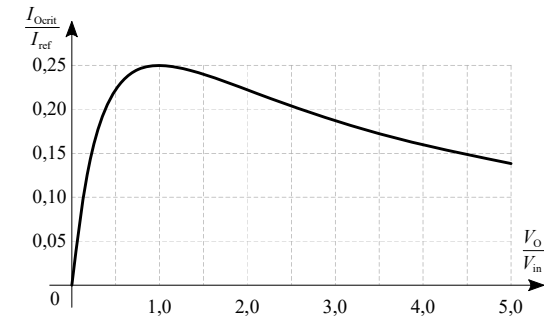
$$I_{Lcrit} = I_{LAVcrit} = \frac{V_{in}DT}{2L}$$

Critical output current I_{Ocrit} is the output current at which the inductor current is equal to I_{Lcrit}

$$I_{Ocrit} = I_{Lcrit} (1-D) = \frac{V_{in}T}{2L} D(1-D)$$

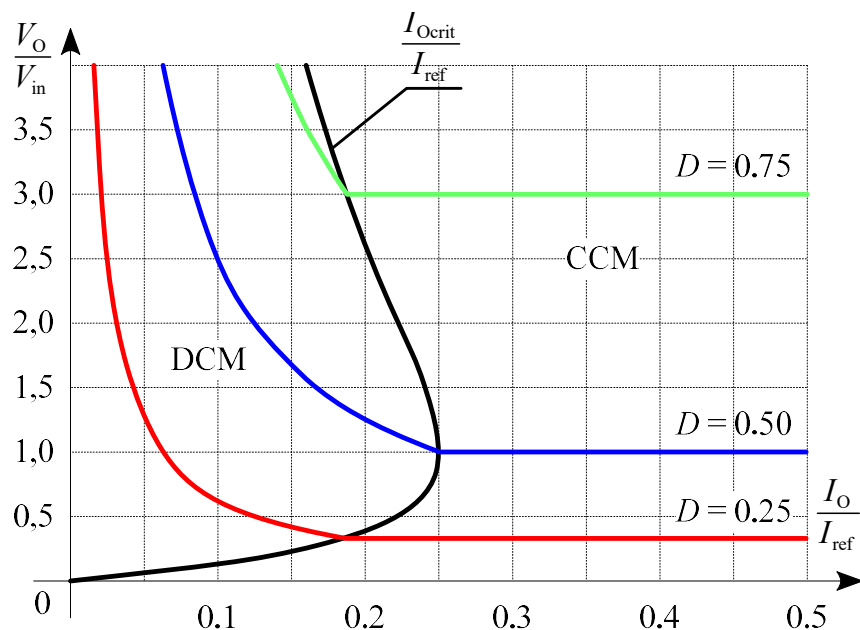
The same as in buck and boost converters

$$\begin{aligned} \frac{V_o}{V_{in}} &= \frac{D}{1-D} = x \Rightarrow D = \frac{x}{x+1} \\ I_{Ocrit} &= \frac{V_{in}T}{2L} \frac{x}{x+1} \left(1 - \frac{x}{x+1}\right) = \\ &= \frac{V_{in}T}{2L} \frac{x}{(x+1)^2} \end{aligned}$$

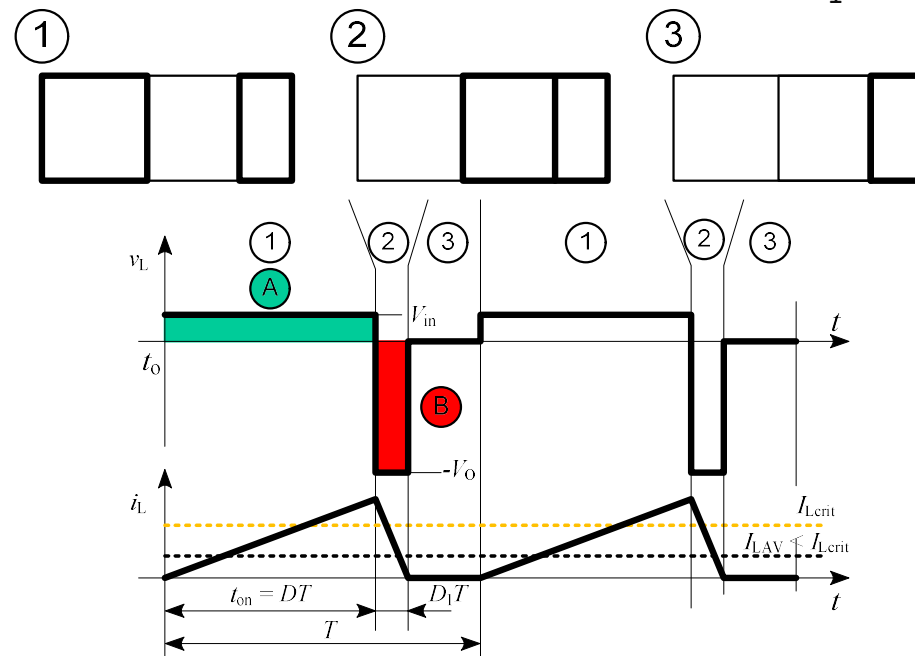


Buck-boost converter

The output characteristic for $V_{in} = \text{const}$



For DCM it is needed to calculate duty cycle D_1



$$V_{LAV} = V_{in}D + (-V_O)D_1 = 0 \Rightarrow D_1 = \frac{V_{in}}{V_O}D \text{ or } D_1 = \frac{1}{\frac{V_O}{V_{in}}}D$$

$$I_O = \frac{1}{2}\Delta I_L D_1 = \frac{1}{2}\frac{V_{in}T}{L}DD_1 = \frac{1}{2}\frac{V_{in}T}{L}D^2\frac{V_{in}}{V_O} \Rightarrow \frac{V_O}{V_{in}} = \frac{1}{2}\frac{V_{in}T}{LI_O}D^2 = D^2\frac{I_{ref}}{I_O}$$

Buck-boost converter

The output characteristic for $V_O = \text{const}$

Defining output reference current: $I_{\text{Oref}} = \frac{V_O T}{2L}$

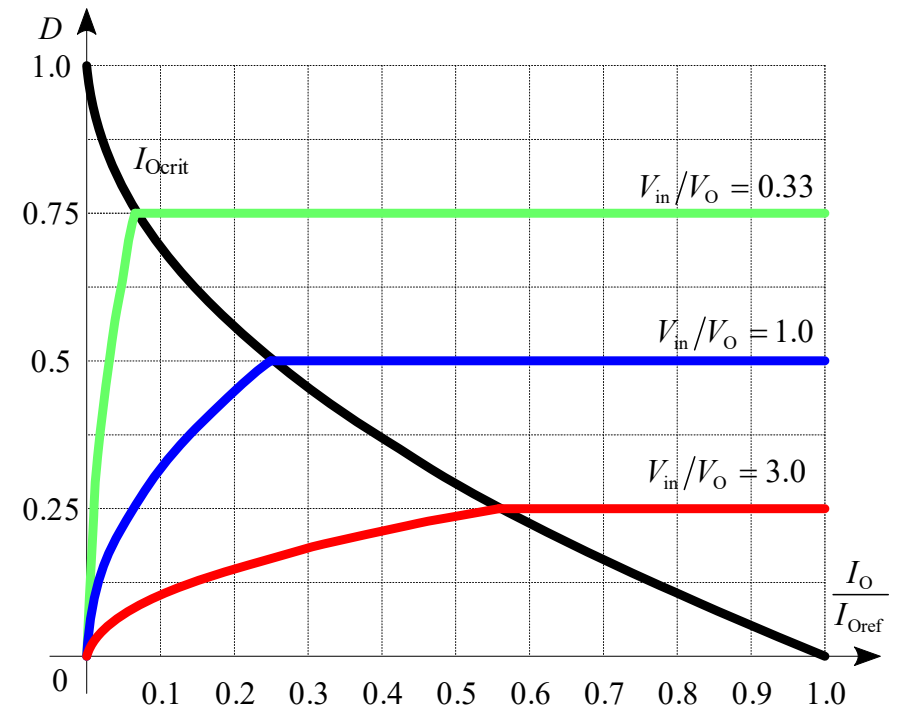
The critical output current is:

$$I_{\text{Ocrit}} = \frac{V_{\text{in}} T}{2L} D(1-D) \stackrel{V_{\text{in}} \frac{D}{1-D} = V_O}{=} \frac{V_O T}{2L} (1-D)^2 = I_{\text{Oref}} (1-D)^2$$

The relation between the output current and duty cycle in DCM is:

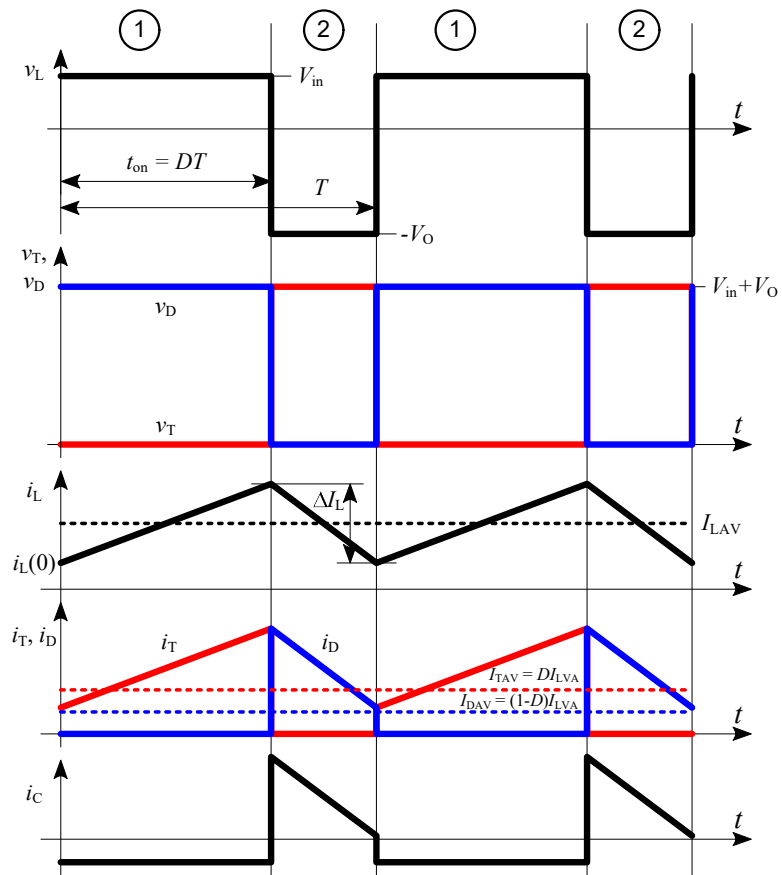
$$I_O = \frac{1}{2} \Delta I_L D_1 = \frac{V_{\text{in}} T}{2L} D D_1 \stackrel{D_1 = \frac{1}{V_O/V_{\text{in}}}}{=} I_{\text{Oref}} D^2 \left(\frac{V_{\text{in}}}{V_O} \right)^2$$

The output characteristic for $V_O = \text{const}$

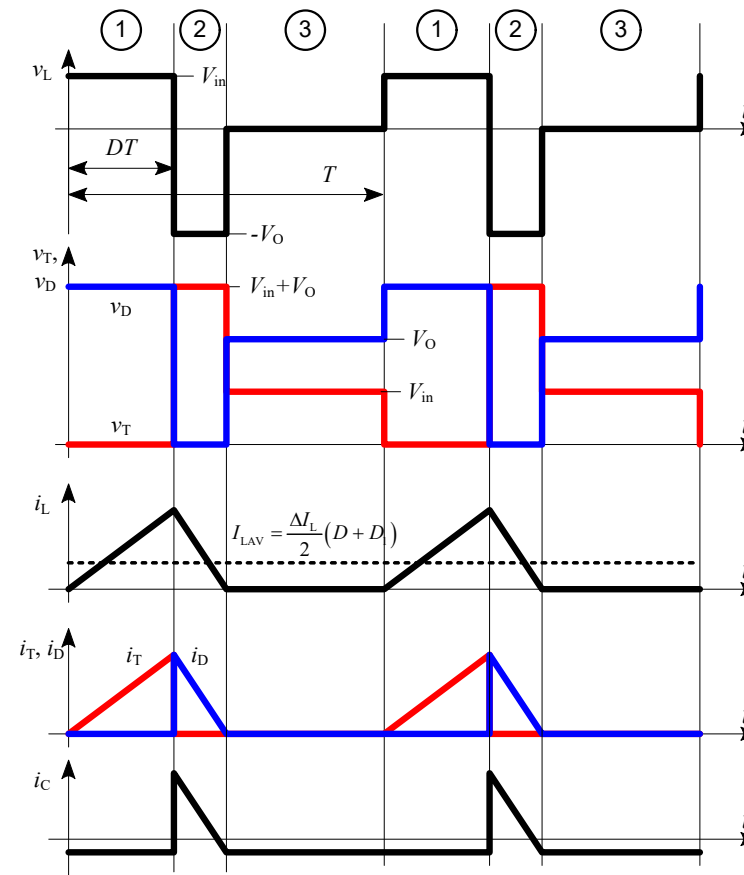


Buck-boost converter

Waveforms in CCM

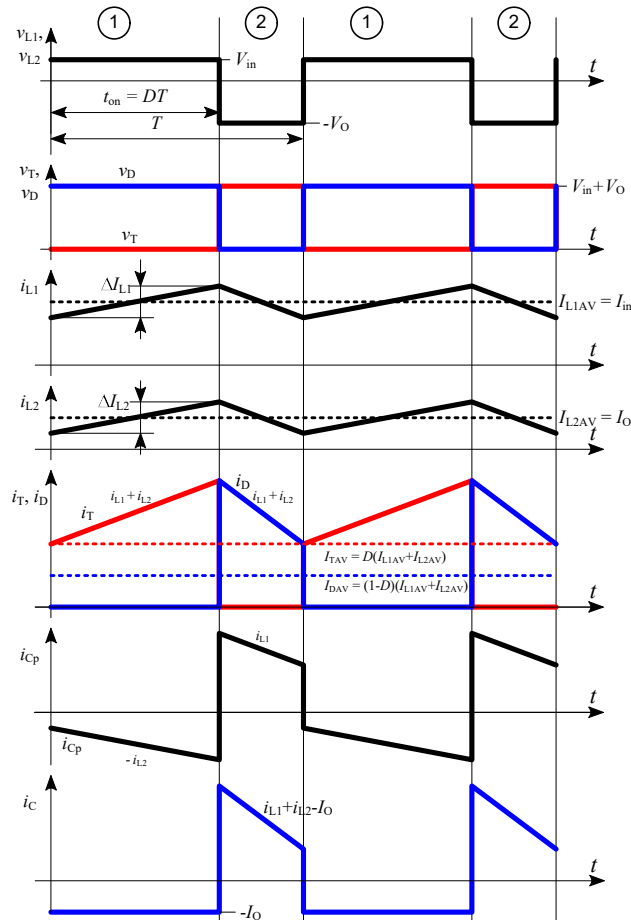


Waveforms in DCM

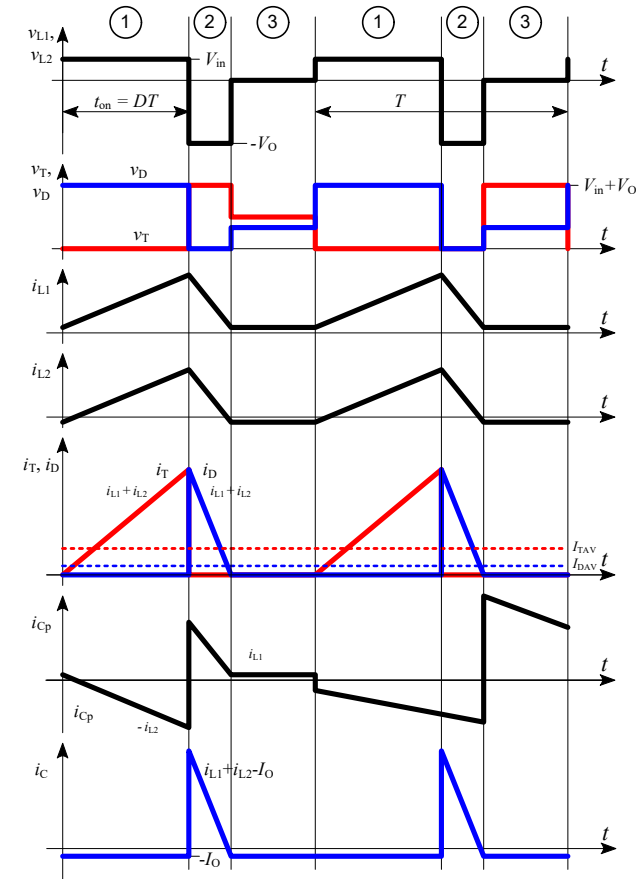


SEPIC converter

Waveforms in CCM

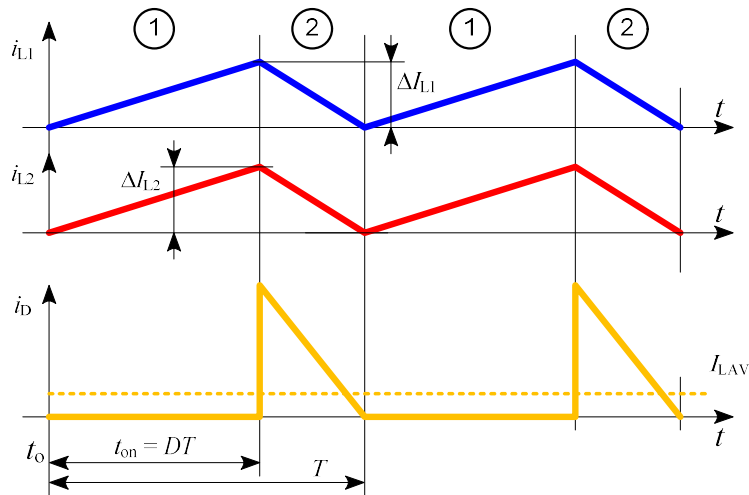
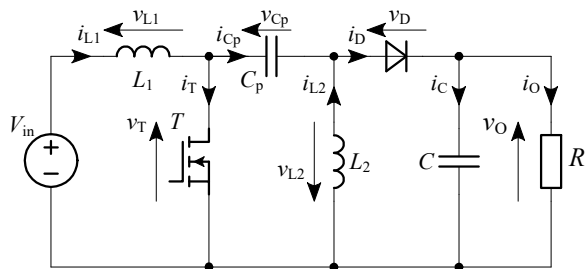


Waveforms in DCM



SEPIC converter

Boundary current occurs when diode current reaches zero value at the end of the switching period



Critical output current I_{Ocrit} is at the boundary between CCM & DCM

$$I_{Ocrit} = \frac{1}{2}(\Delta I_{L1} + \Delta I_{L2})(1-D) = \frac{V_{in}T}{2} \left(\frac{1}{L_1} + \frac{1}{L_2} \right) D(1-D)$$

$$I_{Ocrit} = \frac{V_{in}T}{2L_1} \left(\frac{L_1 + L_2}{L_2} \right) D(1-D) = I_{ref} \left(\frac{L_1 + L_2}{L_2} \right) D(1-D)$$

$$= I_{ref} \left(\frac{L_1 + L_2}{L_2} \right) \frac{\frac{V_o}{V_{in}}}{1 + \frac{V_o}{V_{in}}} \left(\frac{1}{1 + \frac{V_o}{V_{in}}} \right)$$

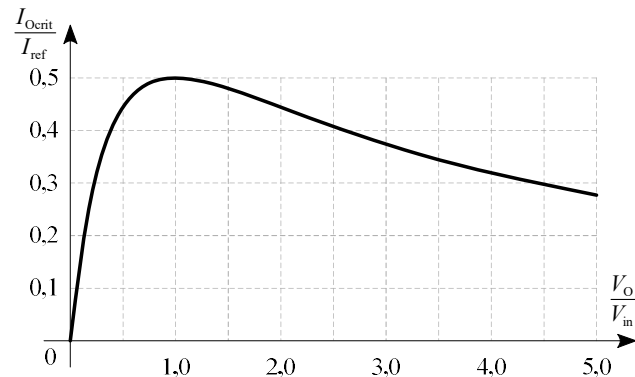
SEPIC converter

Boundary between DCM and CCM in sepic converter can be shown as a characteristic of $V_O/V_{in} = f(I_O)$

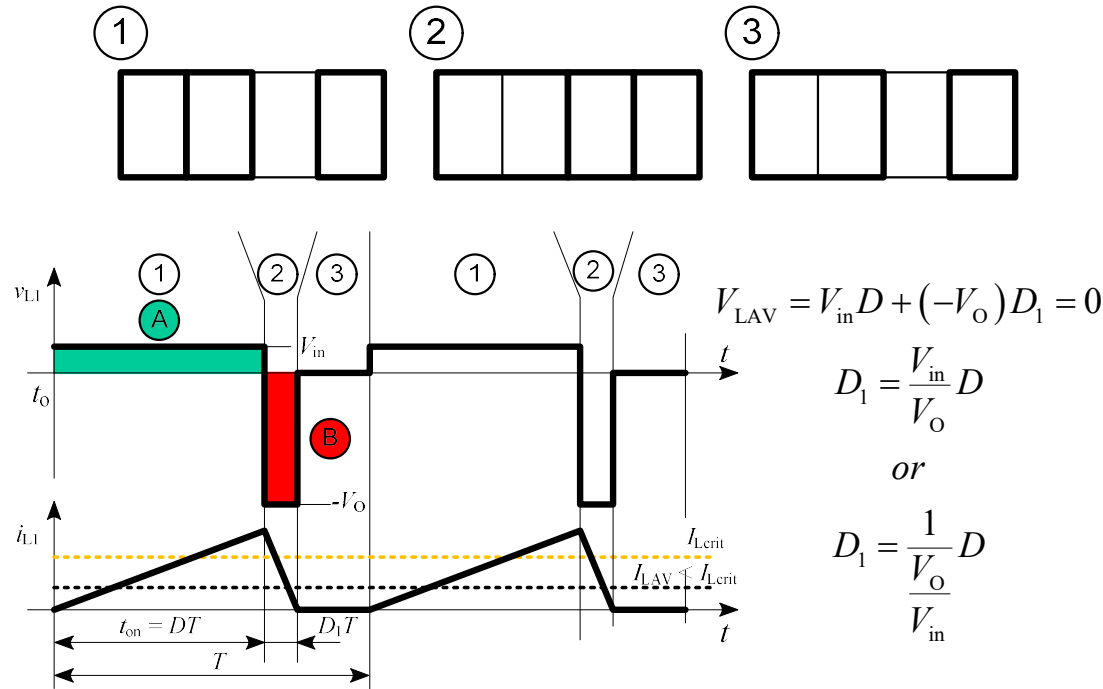
$$I_{Ocrit} = \frac{V_{in}T}{2L_1} \left(\frac{L_1 + L_2}{L_2} \right) D(1-D)$$

$$\frac{V_O}{V_{in}} = \frac{D}{1-D} = x \Rightarrow D = \frac{x}{x+1}$$

$$I_{Ocrit} = \frac{V_{in}T}{2L_1} \left(\frac{L_1 + L_2}{L_2} \right) \frac{x}{x+1} \left(1 - \frac{x}{x+1} \right) = \frac{V_{in}T}{2L_1} \left(\frac{L_1 + L_2}{L_2} \right) \frac{x}{(x+1)^2}$$



For DCM it is needed to calculate duty cycle D_1

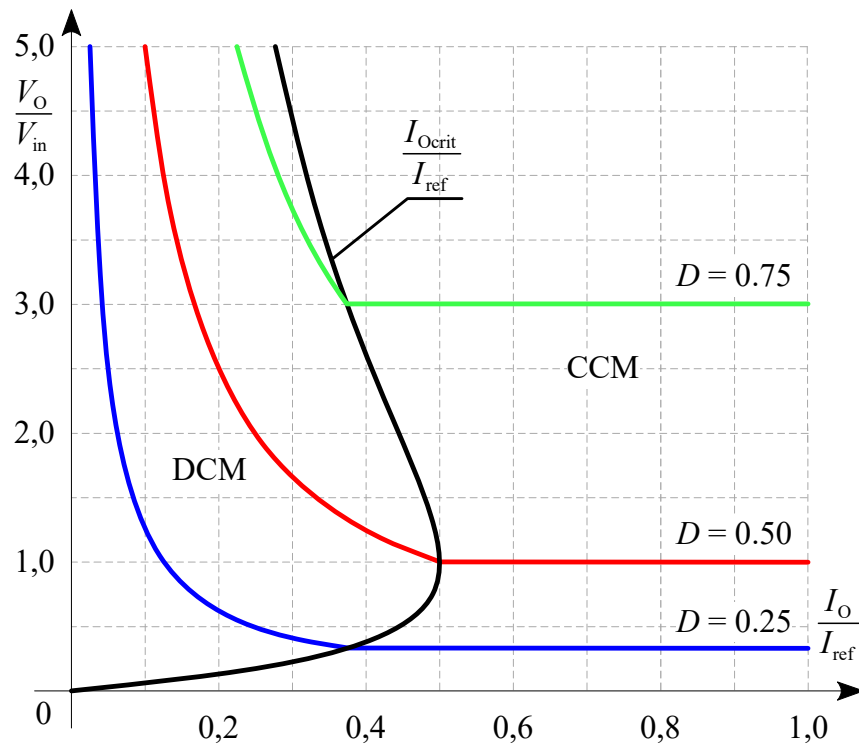


$$I_O = \frac{1}{2}(\Delta I_{L1} + \Delta I_{L2})D_1 = \frac{1}{2} \frac{V_{in}T}{L_1} \left(\frac{L_1 + L_2}{L_2} \right) DD_1 = \frac{1}{2} \frac{V_{in}T}{L_1} \left(\frac{L_1 + L_2}{L_2} \right) D^2 \frac{V_{in}}{V_O}$$

$$\frac{V_O}{V_{in}} = \frac{1}{I_O} \frac{1}{2} \frac{V_{in}T}{L_1} \left(\frac{L_1 + L_2}{L_2} \right) D^2 = D^2 \frac{I_{ref}}{I_O} \left(\frac{L_1 + L_2}{L_2} \right)$$

SEPIC converter

The output characteristic for $V_{in} = \text{const}$

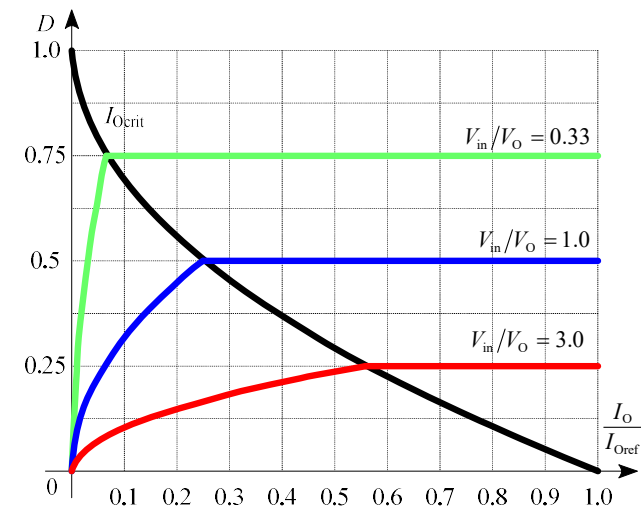


Defining output reference current: $I_{Oref} = \frac{V_O T}{2L_1}$

$$I_{Ocrit} = \frac{V_{in} T}{2L_1} \left(\frac{L_1 + L_2}{L_2} \right) D(1-D) \stackrel{V_{in} \frac{D}{1-D} = V_O}{=} \frac{V_O T}{2L_1} \left(\frac{L_1 + L_2}{L_2} \right) (1-D)^2 =$$

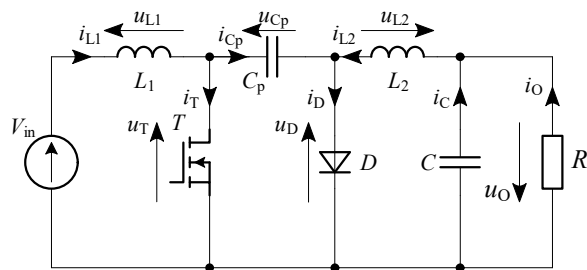
$$= I_{Oref} \left(\frac{L_1 + L_2}{L_2} \right) (1-D)^2$$

$$I_O = \frac{1}{2} \Delta I_L D_1 = \frac{V_{in} T}{2L} D D_1 \stackrel{D_1 = \frac{1}{V_O/V_{in}} D}{=} I_{Oref} D^2 \left(\frac{V_{in}}{V_O} \right)^2$$

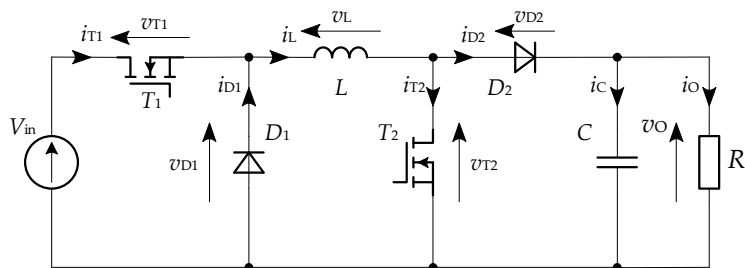


Other Buck-Boost converters

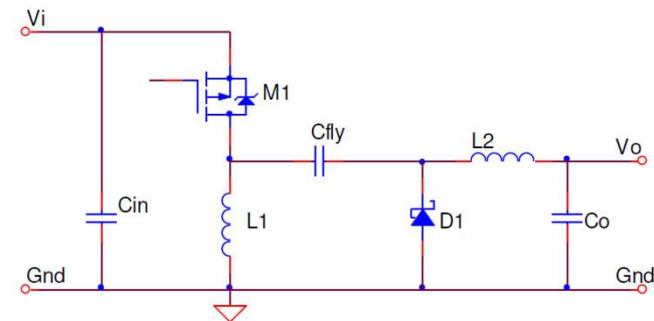
Cuk converter



Two-transistor buck-boost converter



Zeta converter



Interleaved converter